### 68.1 What is an OP-AMP?

It is a very high-gain, high- $r_{i n}$ directly-coupled negative-feedback amplifier which can amplify signals having frequency ranging from $\mathbf{0} \mathbf{H z}$ to a little beyond $\mathbf{1 ~ M H z}$. They are made with different internal configurations in linear ICs. An $O P-A M P$ is so named because it was originally designed to perform mathematical operations like summation, subtraction, multiplication, differentiation and integration etc. in analog computers. Present day usage is much wider in scope but the popular name $O P-A M P$ continues.

Typical uses of $O P-A M P$ are : scale changing, analog computer operations, in instrumentation and control systems and a great variety of phase-shift and oscillator circuits. The $O P-A M P$ is available in three different packages (i) standard dual-in-line package (DIL) (ii) TO-5 case and (iii) the flat-pack.

Although an $O P-A M P$ is a complete amplifier, it is so designed that external components (resistors, capacitors etc.) can be connected to its terminals to change its external characteristics. Hence, it is relatively easy to tailor this amplifier to fit a particular application and it is, in fact, due to this versatility that $O P-A M P s$ have become so popular in industry.

An OP-AMP IC may contain two dozen transistors, a dozen resistors and one or two capacitors.

## Example of OP-AMPs

1. $\mu \mathrm{A} 709$-is a high-gain operational amplifier constructed on a single silicon chip using planar epitaxial process.
It is intended for use in dc servo systems, high-impedance analog computers and in lowlevel instrumentation applications.
It is manufactured by Semiconductors Limited, Pune.
2. [LM 108 - LM 208]— Manufactured by Semiconductors Ltd. Bombay,
3. $C A 741 C T$ and $C A 741 T$-these are high-gain operational amplifiers which are intended for use as (i) comparator, (ii) integrator, (iii) differentiator, (iv) summer, (v) dc amplifier, (vi) multivibrator and (vii) bandpass filter.

Manufactured by Bharat Electronics Ltd (BEL), Bangalore.

### 68.2. OP-AMP Symbol

Standard triangular symbol for an $O P-A M P$ is shown in Fig. 68.1 (a) though the one shown in Fig. 68.1 (b) is also used often. In Fig. 68.1 (b), the common ground line has been omitted. It also does not show other necessary connections such as for dc power and feedback etc.

The $O P-A M P$ ' $s$ input can be single-


Fig. 68.1 ended or double-ended (or differential input) depending on whether input voltage is applied to one input terminal only or to both. Similarly, amplifier's output can also be either single-ended or doubleended. The most common configuration is two input terminals and a single output.

All OP-AMPs have a minimum of five terminals :

1. inverting input terminal,
2. output terminal,
3. negative bias supply terminal.
4. non-inverting input terminal,
5. positive bias supply terminal,

### 68.3. Polarity Conventions

In Fig. 68.1 (b), the input terminals have been marked with minus (-) and plus (+) signs. These are meant to indicate the inverting and noninverting terminals only [Fig. 68.2]. It simply means that a signal applied at negative input terminal will appear amplified but phase-inverted at the output terminal as shown in Fig. 68.2 (b). Similarly, signal applied at the positive input terminal will appear amplified and inphase at the output. Obviously, these plus and minus polarities indicate phase reversal only. It does not mean that voltage $v_{1}$ and $v_{2}$ in Fig. $68.2(a)$ are negative and positive respectively. Additionally, it also does not imply that a positive input
 voltage has to be connected to the plus-marked non-inverting terminal 2 and negative input voltage to the negative-marked inverting terminal 1.
In fact, the amplifier can be used 'either way up' so to speak. It may also be noted that all input and output voltages are referred to a common reference usually the ground shown in Fig. 68.1 (a).

### 68.4. Ideal Operational Amplifier

When an $O P-A M P$ is operated


Fig. 68.2 without connecting any resistor or capacitor from its output to any one of its inputs (i.e., without feedback), it is said to be in the open-loop condition. The word 'open loop' means that feedback path or loop is open. The specifications of $O P-A M P$ under such condition are called open-loop specifications.

An ideal $O P-A M P$ (Fig. 68.3) has the following characteristics :

1. its open-loop gain $A_{v}$ is infinite i.e., $A_{v}=-\infty$
2. its input resistance $R_{i}$ (measured between inverting and non-inverting terminals) is infinite i.e., $R_{i}=\infty \mathrm{ohm}$
3. its output resistance $R_{0}$ (seen looking back into output terminals) is zero i.e., $R_{0}=0 \Omega$
4. it has infinite bandwith i.e., it has flat frequency response from dc to infinity.

Though these characteristics cannot be achieved in practice, yet an ideal $O P-A M P$ serves as a convenient reference against which real $O P-A M P s$ may be evaluated.

Following additional points are worth noting :

1. infinte input resistance means that input current $i=0$ as indicated in Fig. 68.3. It means that an ideal $O P-A M P$ is a voltage-controlled device.
2. $R_{0}=0$ means that $v_{0}$ is not dependent on the load resistance connected across the output.
3. though for an ideal $O P-A M P A_{v}=\infty$, for an actual one, it is extremely high i.e., about $10^{6}$. However, it does not mean that 1 V signal will be amplified to $10^{6} \mathrm{~V}$ at the output. Actually, the maximum value of $v_{0}$ is limited by the basis supply voltage, typically $\pm 15 \mathrm{~V}$. With $A_{v}=10^{6}$ and $v_{0}=15 V_{2}$ the maximum value of input voltage is limited to $15 / 10^{6}=15 \mu \mathrm{~V}$. Though $1 \mu \mathrm{~V}$ in the $O P-A M P$, can cer-


Fig. 68.3 tainly become 1 V .

### 68.5. Virtual Ground and Summing Point

In Fig. 68.4 is shown an $O P-A M P$ which employs negative feedback with the help of resistor $R_{f}$ which feeds a portion of the output to the input.

Since input and feedback currents are algebraically added at point $A$, it is called the summing point.

The concept of virtual ground arises from the fact that input voltage $v_{i}$ at the inverting terminal of the $O P-A M P$ is forced to such a small value that, for all practical purposes, it may be assumed to be zero. Hence, point $A$ is essentially at ground voltage and is referred to as virtual ground. Obviously, it is not the actual ground, which, as seen from


Fig. 68.4 Fig. 68.4, is situated below.

### 68.6. Why $\mathrm{V}_{i}$ is Reduced to Almost Zero ?

When $v_{1}$ is applied, point $A$ attains some positive potential and at the same time $v_{0}$ is brought into existence. Due to negative feedback, some fraction of the output voltage is fed back to point $A$ antiphase with the voltage already existing there (due to $v_{1}$ ).

The algebraic sum of the two voltages is almost zero so that $v_{i} \cong 0$. Obviously, $v_{i}$ will become exactly zero when negative feedback voltage at $\mathbf{A}$ is exactly equal to the positive voltage produced by $\mathrm{v}_{1}$ at A .

Another point worth considering is that there exists a virtual short between the two terminals of the $O P$-AMP because $v_{i}=0$. It is virtual because no current flows (remember $i=0$ ) despite the existence of this short.

### 68.7. OP-AMP Applications

We will consider the following applications :

1. as scalar or linear (i.e., small-signal) constant-gain amplifier both inverting and non-inverting,
2. as unity follower,
3. Subtractor,
4. Differentiator

Now, we will discuss the above circuits one by one assuming an ideal $O P$ AMP.

### 68.8. Linear Amplifier

We will consider the functioning of an $O P-A M P$ as constant-gain amplifier both in the inverting and non-inverting configurations.
(a) Inverting Amplifier or Negative Scale.

As shown in Fig. 68.5, noninverting
3. Adder or Summer
5. Integrator
7. Comparator.


Fig. 68.5
terminal has been grounded, whereas $R_{1}$ connects the input signal $v_{1}$ to the inverting input. A feedback resistor $R_{f}$ has been connected from the output to the inverting input.

Gain
Since point $A$ is at ground potential ${ }^{*}, i_{1}=\frac{v_{\text {in }}}{R_{1}}=\frac{v_{1}}{R_{1}}$

$$
i_{2}=\frac{-v_{0}}{R_{f}} \text { Please note }- \text { ve sign }
$$

Using $K C L$ (Art. 2.2) for point $A$,

$$
\begin{array}{lllll} 
& i_{1}+\left(-i_{2}\right) & =0 \quad \text { or } \quad \frac{v_{1}}{R_{1}}+\frac{v_{0}}{R_{f}}=0 \quad \text { or } \quad \frac{v_{0}}{R_{f}}=-\frac{v_{1}}{R_{1}} \quad \text { or } \quad \frac{v_{0}}{v_{1}}=-\frac{R_{f}}{R_{1}} \\
\therefore & A_{v} & =-\frac{R_{f}}{R_{1}} \text { or } A_{v}=-K & \text { Also, } v_{0}=-K v_{i n}
\end{array}
$$

It is seen from above, that closed-loop gain of the inverting amplifier depends on the ratio of the two external resistors $R_{1}$ and $R_{f}$ and is independent of the amplifier parameters.

It is also seen that the $O P-A M P$ works as a negative scaler. It scales the input i.e., it multiplies the input by a minus constant factor $K$.
(b) Non-inverting Amplifier or Positive Scaler

This circuit is used when there is need for an output which is equal to the input multiplied by a positive constant. Such a positive scaler circuit which uses negative feedback but provides an output that equals the input multiplied by a positive constant is shown in Fig. 68.6.

Since input voltage $v_{2}$ is applied to the non-inverting terminal, the circuits is also called non-inverting amplifier.


Fig. 68.6

Here, polarity of $v_{0}$ is the same as that $v_{2}$ i.e., both are positive.
Gain
Because of virtual short between the two $O P-A M P$ terminals, voltage across $R_{1}$ is the input voltage $v_{2}$. Also, $v_{0}$ is applied across the series combination of $R_{1}$ and $R_{f}$

$$
\begin{array}{ll}
\therefore & v_{i n}=v_{2}=i R_{1}, v_{0}=i\left(R_{1}+R_{f}\right) \\
\therefore & A_{v}=\frac{v_{0}}{v_{i n}}=\frac{i\left(R_{1}+R_{f}\right)}{i R_{1}} \quad \text { or } A_{v}=\frac{R_{1}+R_{f}}{R_{1}}=\left(1+\frac{R_{f}}{R_{1}}\right)
\end{array}
$$

## Alternative Derivation

As shown in Fig. 68.7, let the currents through the two resistors be $i_{1}$ and $i_{2}$.

The voltage across $R_{1}$ is $v_{2}$ and that across $R_{f}$ is $\left(v_{0}-v_{2}\right)$.

$$
\therefore \quad i_{1}=\frac{v_{2}}{R_{1}} \text { and } i_{2}=\frac{v_{0}-v_{2}}{R_{f}}
$$

Applying $K C L$ to junction $A$, we have


Fig. 68.7

* If not, then $i_{1}=\frac{v_{1}-\overline{v_{i}}}{R_{1}}$ and $i_{2}=\frac{v_{0}-v_{1}}{R_{f}}$

$$
\begin{array}{ll} 
& \left(-i_{1}\right)+i_{2}=0 \text { or } \frac{v_{2}}{R_{1}}+\frac{\left(v_{0}-v_{2}\right)}{R_{f}}=0 \\
\therefore & \frac{v_{0}}{R_{f}}=v_{2}\left(\frac{1}{R_{1}}+\frac{1}{R_{f}}\right)=v_{2} \frac{R_{1}+R_{f}}{R_{1} R_{f}} \\
\therefore \quad & \frac{v_{0}}{v_{2}}=\frac{R_{1}+R_{f}}{R_{1}} \text { or } A_{v}=1+\frac{R_{f}}{R_{1}} \quad \text {-as before }
\end{array}
$$

Example 68.1. For the inverting amplifier of Fig. 68.5, $R_{1}=1 \mathrm{~K}$ and $R_{f}=1 \mathrm{M}$. Assuming an ideal OP-AMP amplifier, determine the following circuit values :
(a) voltage gain,
(b) input resistance,
(c) output resistance

Solution. It should be noted that we will be calculating values of the circuit and not for the $O P$ AMP proper.
(a)

$$
A_{\mathrm{v}}=-\frac{R_{f}}{R_{1}}=-\frac{1000 \mathrm{~K}}{1 K}=-1000
$$

(b) Because of virtual ground at $A, R_{\text {in }}=R_{1}=1 \mathrm{~K}$
(c) Output resistance of the circuit equal the output resistance of the $O P-A M P$ i.e., zero ohm.

### 68.9. Unity Follower

It provides a gain of unity without any phase reversal. It is very much similar to the emitter follower (Art 68.8) except that its gain is very much closer to being exactly unity.


Fig. 68.8

This circuit (Fig. 68.8) is useful as a buffer or isolation amplifier because it allows, input voltage $v_{\text {in }}$ to be transferred as output voltage $v_{0}$ while at the same time preventing load resistance $R_{L}$ from loading down the input source. It is due to the fact that its $R_{i}=\infty$ and $R_{0}=0$.

In fact, circuit of Fig. 68.8 can be obtained from that of Fig. 68.6 by putting

$$
R_{1}=R_{f}=0
$$

### 68.10. Adder to Summer

The adder circuit provides an output voltage proportional to or equal to the algebraic sum of two or more input voltages each multiplied by a constant gain factor. It is basically similar to a scaler (Fig. 68.5) except that it has more than one input. Fig. 68.9 shows a three-input inverting adder circuit. As seen, the output voltage is phase-inverted.

## Calculations

As before, we will treat point $A$ as virtual ground

$$
\begin{aligned}
& i_{1}=\frac{v_{1}}{R_{1}} \quad \text { and } \quad i_{2}=\frac{v_{2}}{R_{2}} \\
& i_{3}=\frac{v_{3}}{R_{3}} \quad \text { and } \quad i=-\frac{v_{0}}{R_{f}}
\end{aligned}
$$

Applying $K C I$ to point $A$, we have

$$
i_{1}+i_{2}+i_{3}+(-i)=0
$$



Fig. 68.9
or

$$
\frac{v_{1}}{R_{1}}+\frac{v_{2}}{R_{2}}+\frac{v_{3}}{R_{3}}-\left(\frac{-v_{0}}{R_{f}}\right)=0
$$

$$
\therefore \quad \mathrm{v}_{0}=-\left(\frac{R_{f}}{R_{1}} \mathrm{v}_{1}+\frac{R_{f}}{R_{2}} \mathrm{v}_{2}+\frac{R_{f}}{R_{3}} \mathrm{v}_{3}\right)
$$

$$
\text { or } \quad v_{0}=-\left(K_{1} v_{1}+K_{2} v_{2}+K_{3} v_{3}\right)
$$

The overall negative sign is unavoidable because we are using the inverting input terminal.
If

$$
\begin{aligned}
R_{1} & =R_{2}=R_{3}=R, \text { then } \\
v_{0} & =-\frac{R_{f}}{R}\left(v_{1}+v_{2}+v_{3}\right)=-K\left(v_{1}+v_{2}+v_{3}\right)
\end{aligned}
$$

Hence, output voltage is proportional to (not equal to) the algebraic sum of the three input voltages.

If

$$
R_{f}=R \text {, then ouput exactly equals the sum of inputs. However, if } R_{f}=R / 3
$$

then

$$
v_{0}=-\frac{R / 3}{R}\left(v_{1}+v_{2}+v_{3}\right)=-\frac{1}{3}\left(v_{1}+v_{2}+v_{3}\right)
$$

Obviously, the output is equal to the average of the three inputs.

### 68.11. Subtractor

The function of a subtractor is to provide an output proportional to or equal to the difference of two input signals. As shown in Fig. 68.10 we have to apply the inputs at the inverting as well as noninverting terminals.

## Calculations

According to Superposition Theorem (Art. 2.17) $v_{0}=v_{0}{ }^{\prime}+v_{0}{ }^{\prime \prime}$
where $v_{0}{ }^{\prime}$ is the output produced by $v_{1}$ and $v_{0}{ }^{\prime \prime}$ is that produced by $v_{2}$.
Now, $\quad v_{0}{ }^{\prime}=-\frac{R_{f}}{R_{1}} \cdot v_{1} \quad$...Art 67.37 (a)

$$
\begin{aligned}
& \qquad \mathrm{v}_{0}^{\prime \prime}=\left(1+\frac{R_{f}}{R_{1}}\right) \mathrm{v}_{2} \quad \ldots \text { Art } 67.37(b) \\
& \therefore \quad \mathrm{v}_{0}=\left(1+\frac{R_{f}}{R_{1}}\right) \mathrm{v}_{2}-\frac{R_{f}}{R_{1}} \cdot \mathrm{v}_{1} \\
& \text { Since } R_{f} \gg R_{1} \text { and } R_{f} / R_{1} \gg 1 \text {, hence } \\
& \mathrm{v}_{0} \cong \frac{R_{f}}{R}\left(\mathrm{v}_{2}-\mathrm{v}_{1}\right)=K\left(\mathrm{v}_{2}-\mathrm{v}_{1}\right)
\end{aligned}
$$

$$
\mathrm{v}_{0} \cong \frac{R_{f}}{R_{1}}\left(\mathrm{v}_{2}-\mathrm{v}_{1}\right)=K\left(\mathrm{v}_{2}-\mathrm{v}_{1}\right)
$$

Further, If $R_{f}=R_{1}$, then

$$
v_{0}=\left(v_{2}-v_{1}\right)=\text { difference of the two input voltages }
$$

Obviously, if $R_{f} \neq R_{1}$, then a scale factor is introduced.
Example 68.2. Find the output voltages of an OP-AMP inverting adder for the following sets of input voltages and resistors. In all cases, $R_{f}=1 \mathrm{M}$.

$$
v_{1}=-3 V, v_{2}=+3 V, v_{3}=+2 V ; R_{1}=250 \mathrm{~K}, R_{2}=500 \mathrm{~K}, R_{3}=1 \mathrm{M}
$$

[Electronic Engg. Nagpur Univ. 1991]
Solution. $\quad v_{0}=-\left(K_{1} v_{1}+K_{2} v_{2}+K_{3} v_{3}\right)$

$$
K_{1}=\frac{R_{f}}{R_{1}}=\frac{1000 K}{250 K}=4, K_{2}=\frac{1000}{500}=2, K_{3}=\frac{1 M}{1 M}=1
$$

